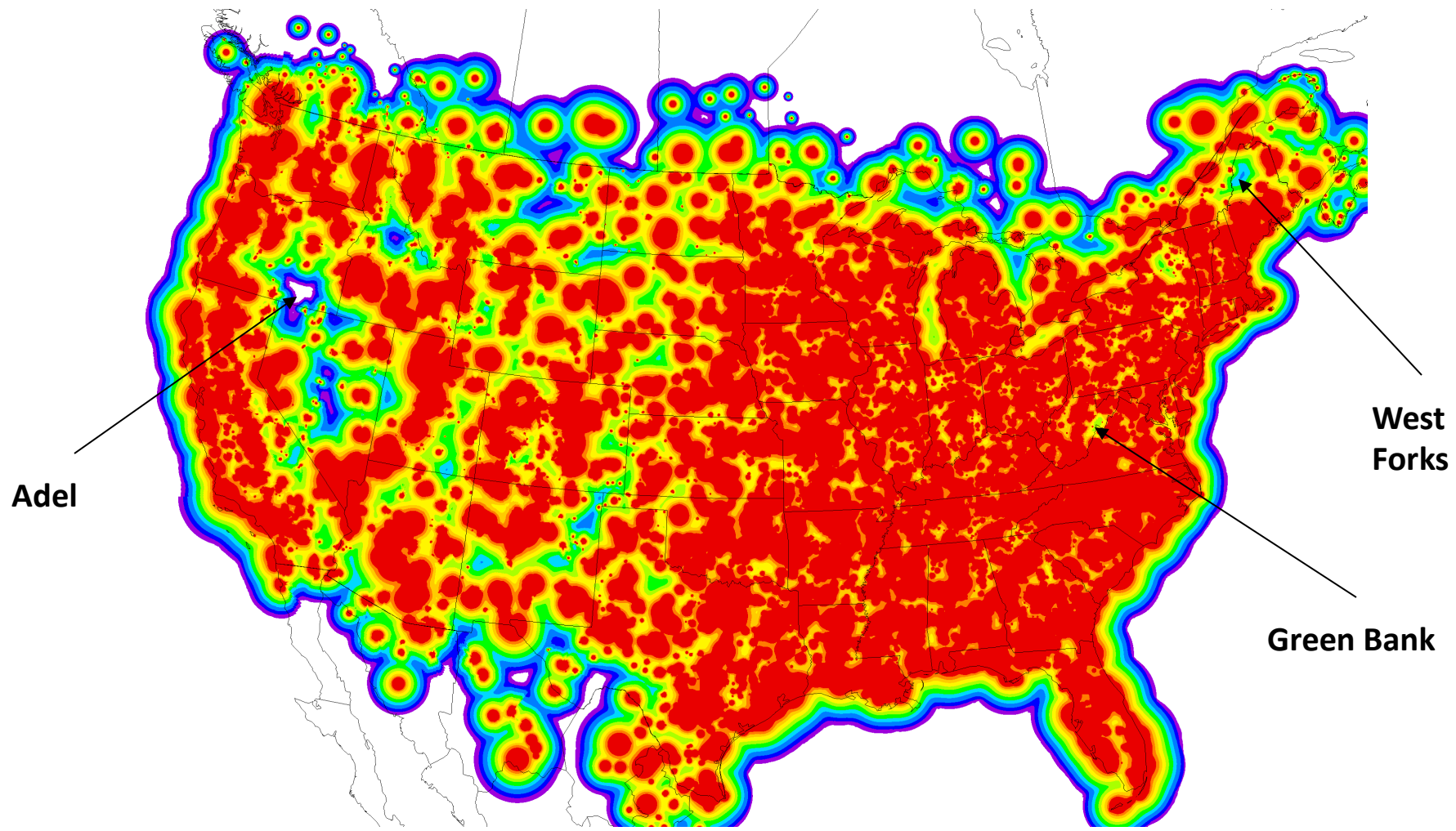


UPDATE on EDGES-3

Alan E. E. Rogers MIT Haystack Observatory
John P. Barrett MIT Haystack Observatory

What I plan to cover

- Location of site in Oregon
- Details of EDGES-3 electronics
- EDGES-3 test in Oregon
- Results from Oregon
- RFI from micrometeors in Oregon, Nevada and the MRO
- Details of Calibration
- Refinements in RFI excision
- Test on soil without ground plane wires
- Tests with smoothing in frequency and Further Refinements in RFI excision



Integrated strength of FM radio for continental USA from radio-locator.com

Quietest regions: West Forks, ME Adel, Oregon -better than Green Bank



“Catlow Valley” region of Oregon first explored in 2009 as a potential EDGES site

Roaring Spring Ranch was contacted in December 2018 and Skull Creek was visited by Mark Derome in January



Terrain surrounding EDGES-3 location near Skull Creek Reservoir



Echart Creek

Skull Creek

Skull Creek Rd

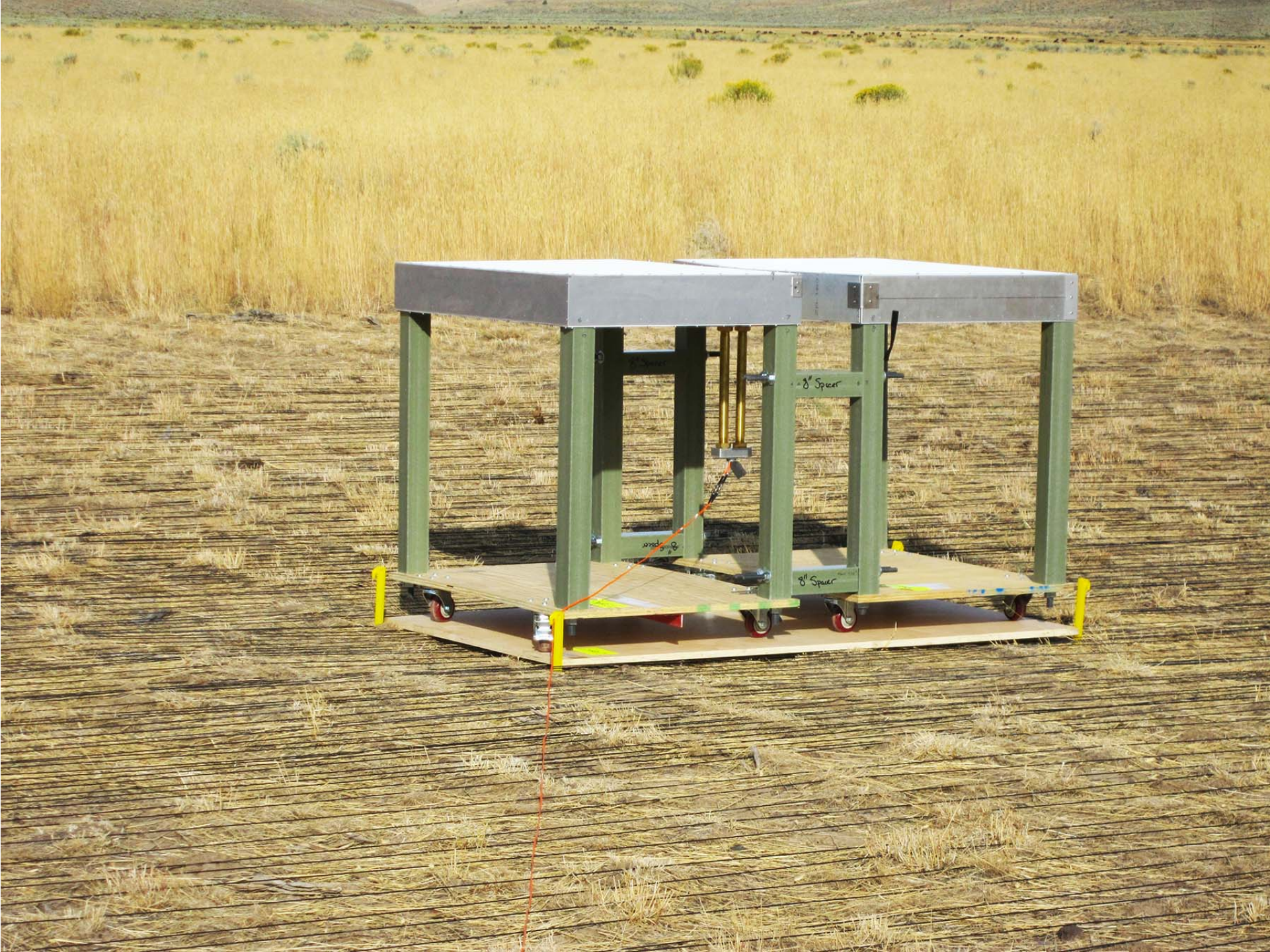
Skull Creek Rd

Skull Creek Reservoir

Cattle Branding Area

EDGES-3

Google



8" Spacer

8" Spacer

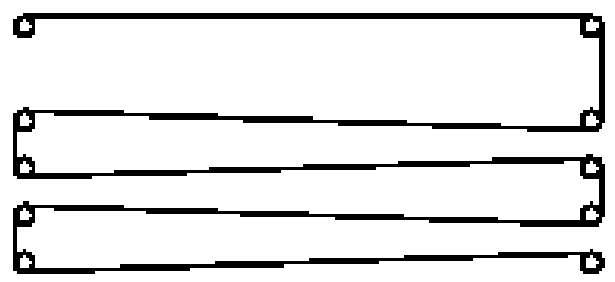
18
10
6
0

48 pegs spaced
12.5 cm apart
i.e. 8 per meter

pegs
12.5cm
apart

magnified illustration

magnified illustration
of change in peg spacing



64 pegs spaced
6.25 cm apart
i.e. 16 per meter

pegs
12.5cm
apart

Wire meanders
around pegs and
goes back and forth
with spacing at antenna
which is close to being
equally spaced

48 pegs spaced
12.5 cm apart
i.e. 8 per meter

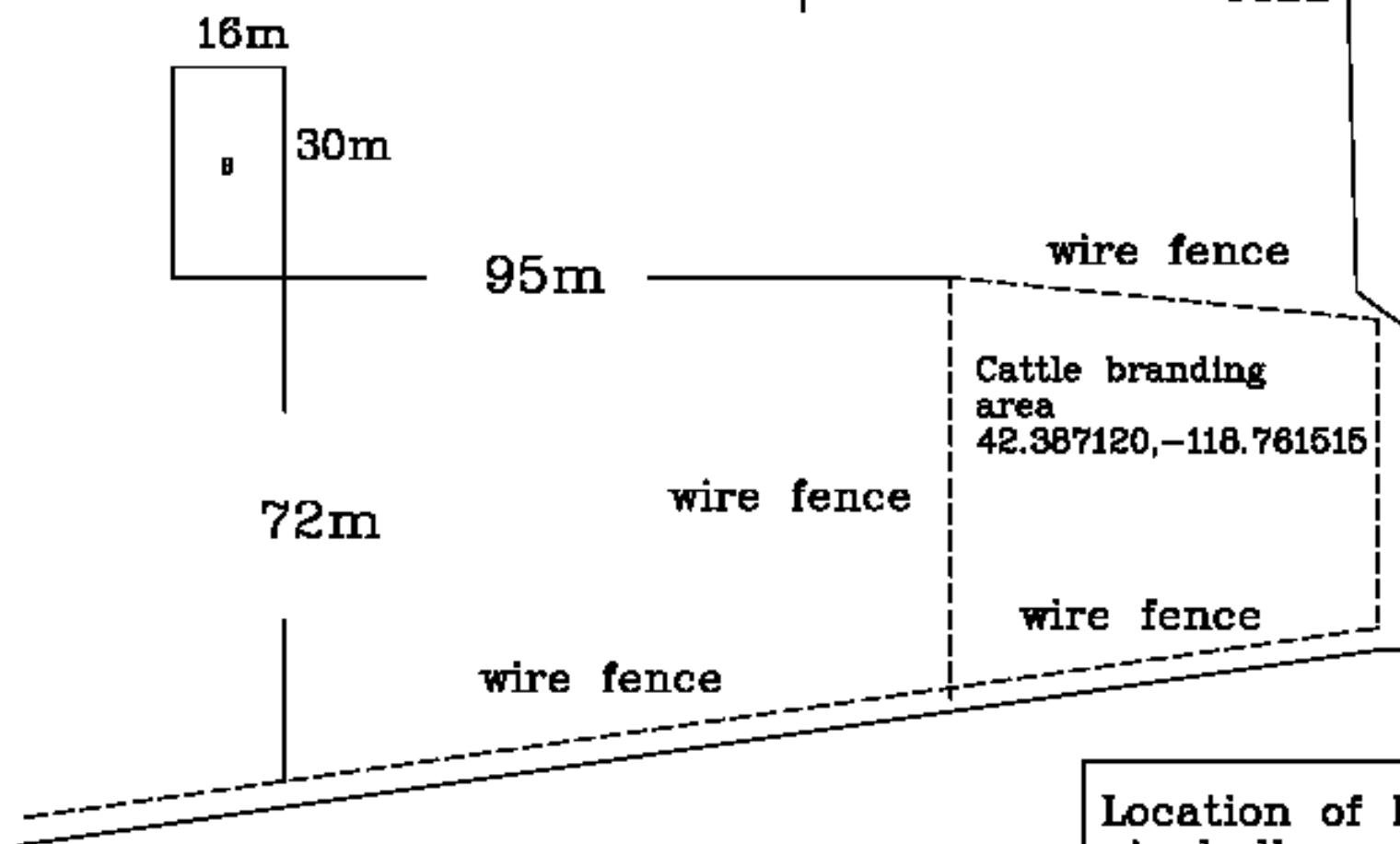
pegs
12.5cm
apart

Wire grid ground plane

EDGES-3 42.387461,-118.76399
on wire grid ground plane

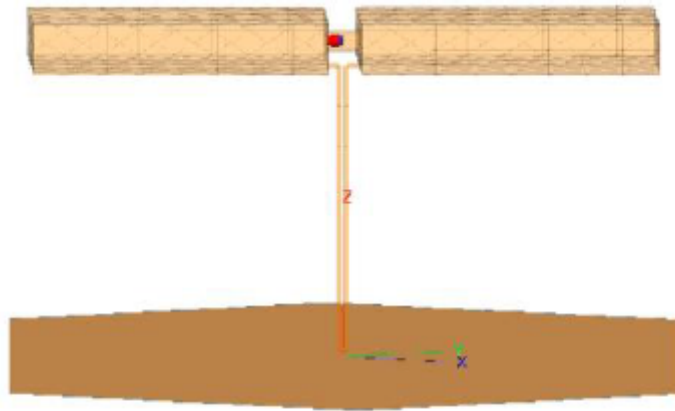
↑ North
+/- 1 deg

farm
road



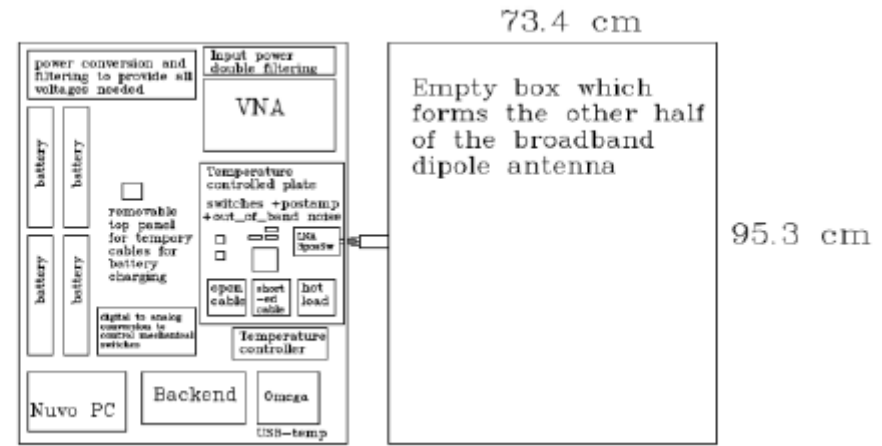
Cattle branding
area
42.387120,-118.761515

Location of EDGES-3
at skull creek
Roaring Springs
Ranch Oregon
aeer 23sep19

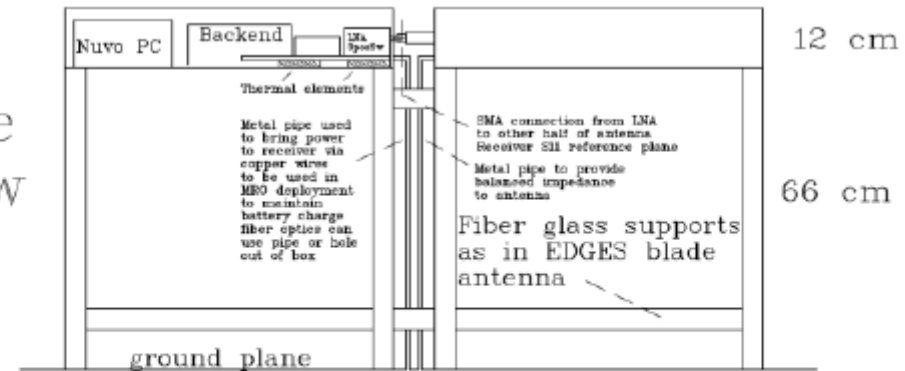


Horizontal profile view of the box-blade antenna design with same dimensions as the planar mid-band blade antenna, except 12cm panel height so that receiver can be located inside an antenna panel.

Top View



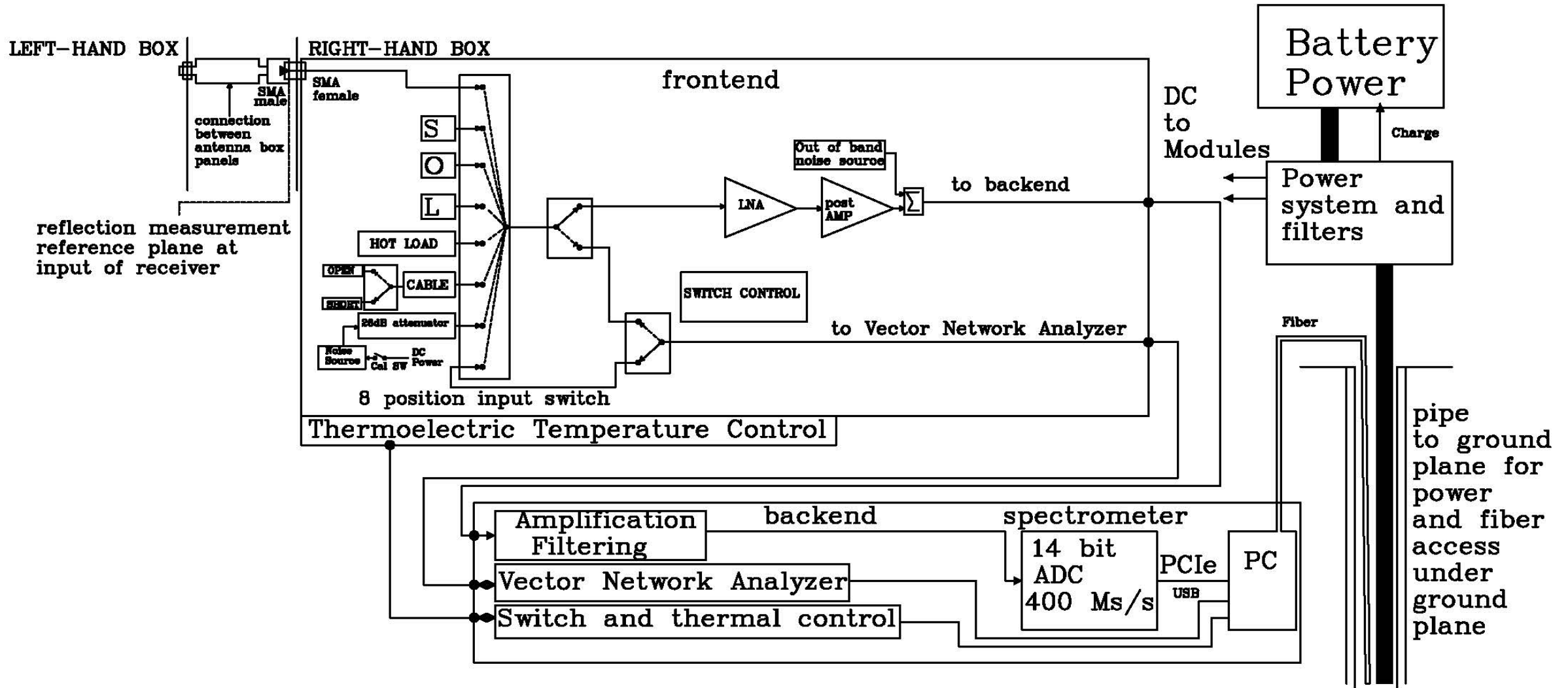
Side View



Layout of components inside antenna panel. Shown with optional batteries for field campaigns.

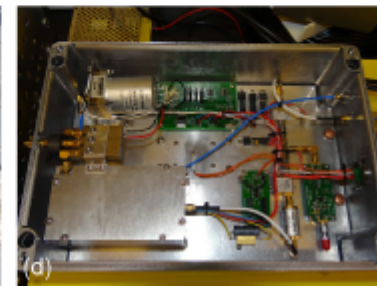
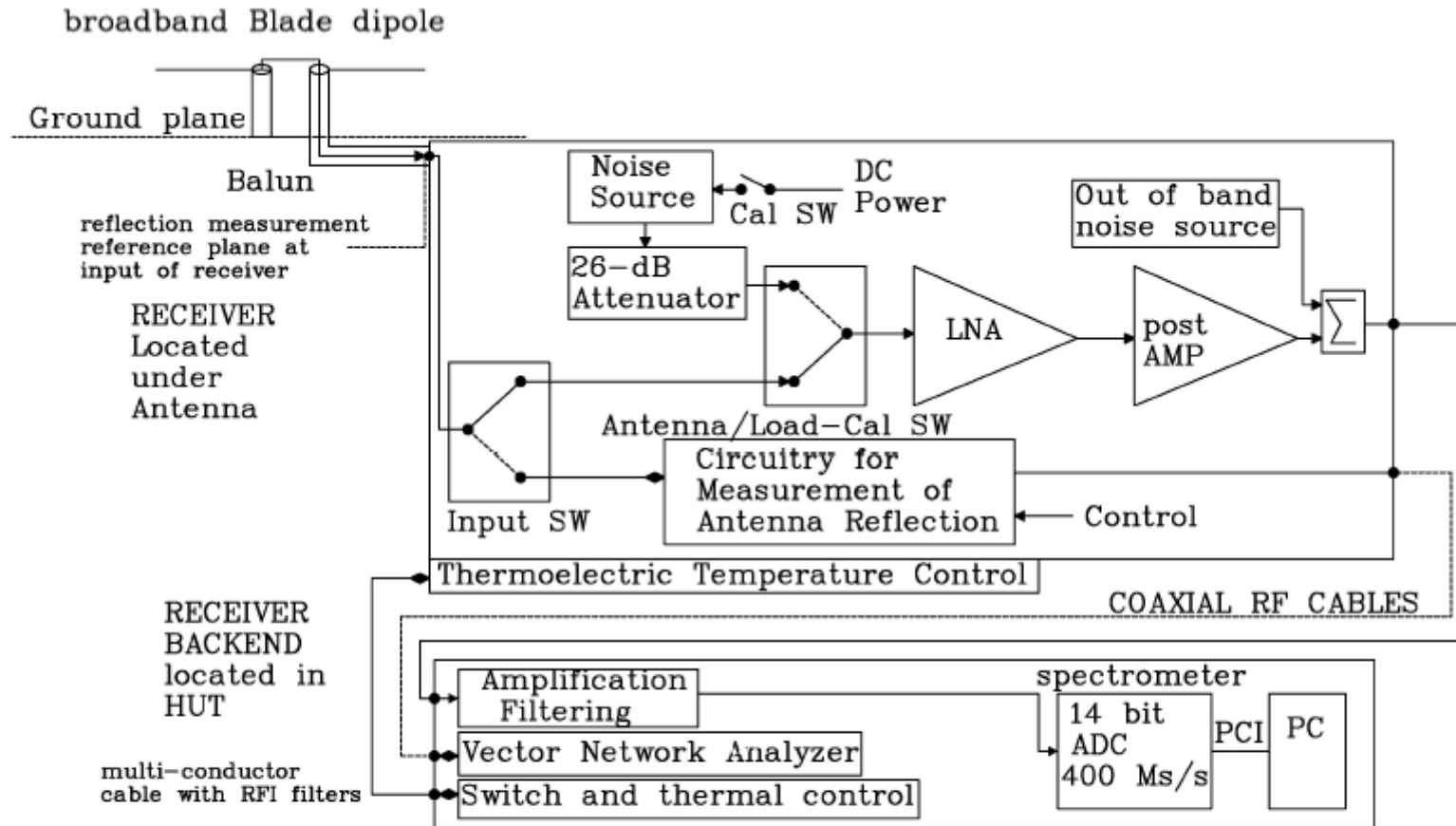
EDGES 3 system with built-in Calibration and all electronics in antenna proposed in November 2018. The advantages are:

- **Reduced loss and less delay in antenna S11 from 22ns for EDGES-2 midband to 14ns for EDGES-3 – since balun is not needed**
- **Automated Calibration**
- **Easier deployment**



Block diagram of EDGES 3 system with built-in Calibration and all electronics in antenna

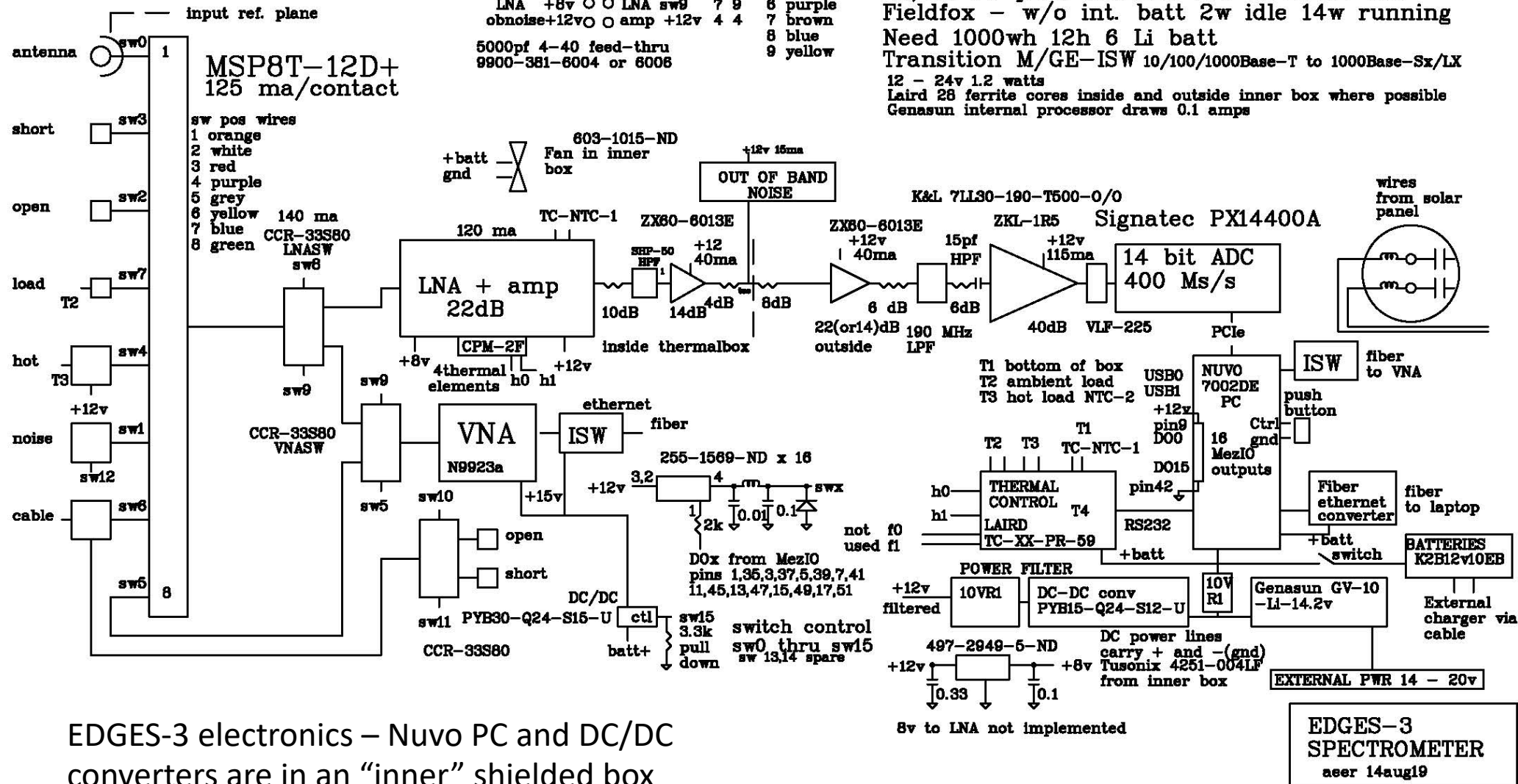
EDGES-2 for comparison requires backend to be located ~ 100m from antenna and manual calibration of receiver in Lab



Power Amps + switches 5w hot 3w
 NUVO 25w idle 72w fastspec
 thermal 10w All max ~96w
 Hot load resistor 74 ohms
 3 ISW 3.6w
 Fan 1.92w

	pin	9-pin colors
VNASWgnd	3	1 grey
VNA sw5	5	2 orange
VNA sw9	9	3 black
op/sh gnd	3	4 red
short sw10	1	5 green
noise+12v sw12	2	6 purple
open sw11	6	7 brown
LNASW gnd	3	8 blue
LNA sw8	8	9 yellow
LNA +8v	7	
LNA sw9	9	
obnoise+12v	4	
amp +12v	4	

Software PC Linux
 Nuvo 32C idle 43C fastspec
 Thermal RS232 Linux
 also monitors temps (100,101,102,104) and voltage(150)
 DC/DC PC pr59 backend ISW in inner box
 Fieldfox - w/o int. batt 2w idle 14w running
 Need 1000wh 12h 6 Li batt
 Transition M/GE-ISW 10/100/1000Base-T to 1000Base-Sx/LX
 12 - 24v 1.2 watts
 Laird 28 ferrite cores inside and outside inner box where possible
 Genasun internal processor draws 0.1 amps

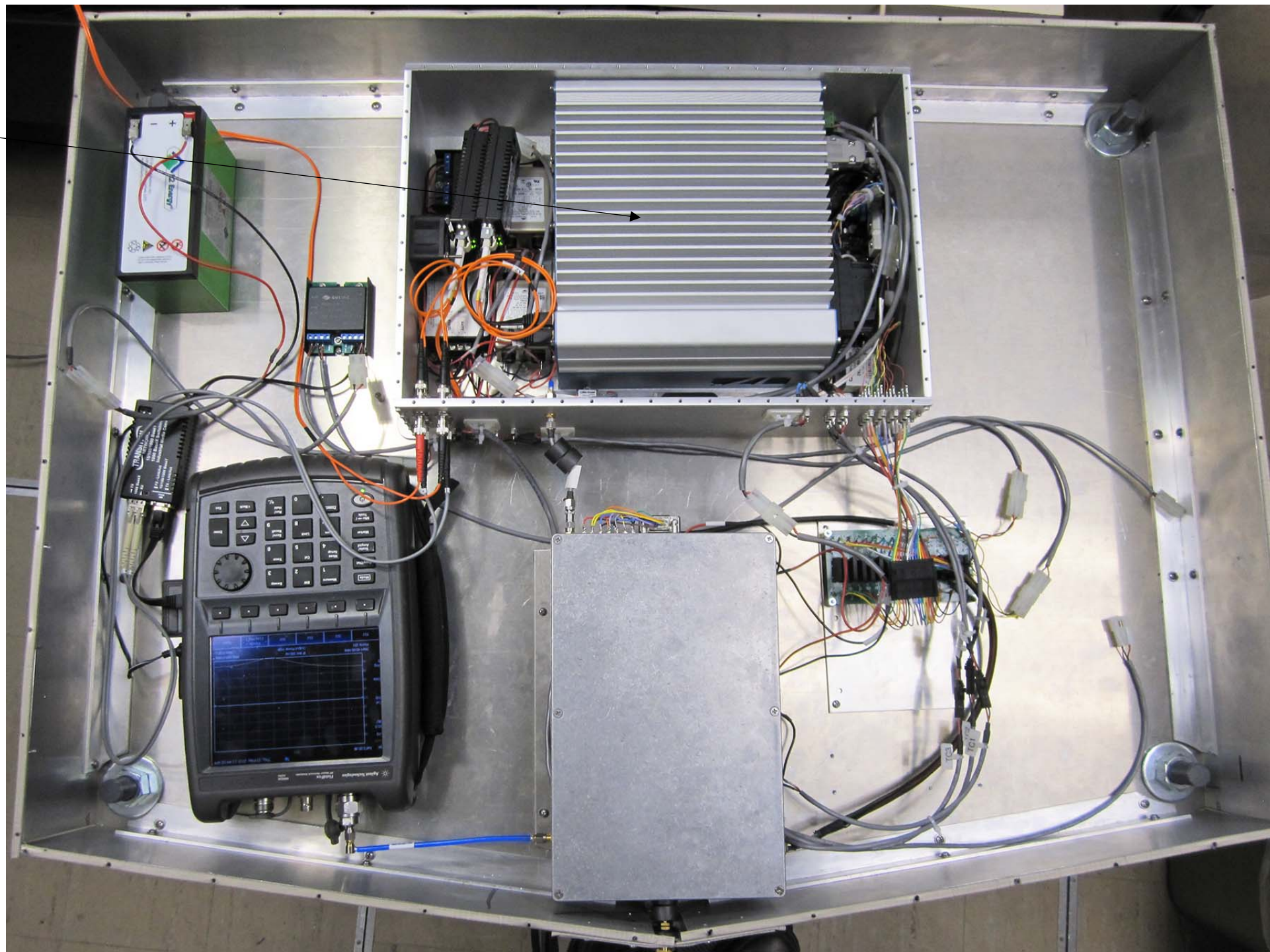


EDGES-3 electronics – Nuvo PC and DC/DC converters are in an “inner” shielded box

“inner” shielded box with cover removed

Notes:
VNA is only turned on for S11 measurements

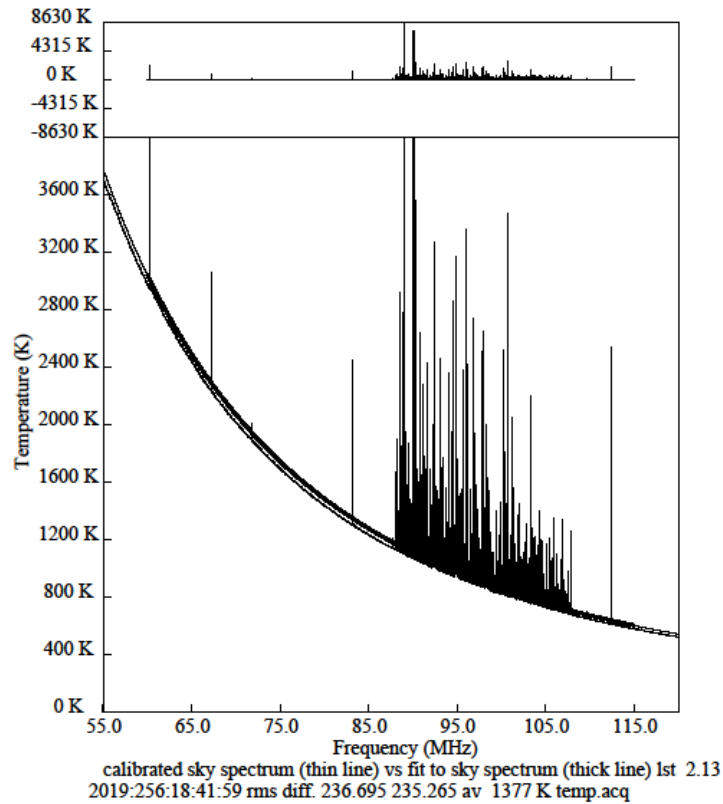
Analog electronics is temperature controlled



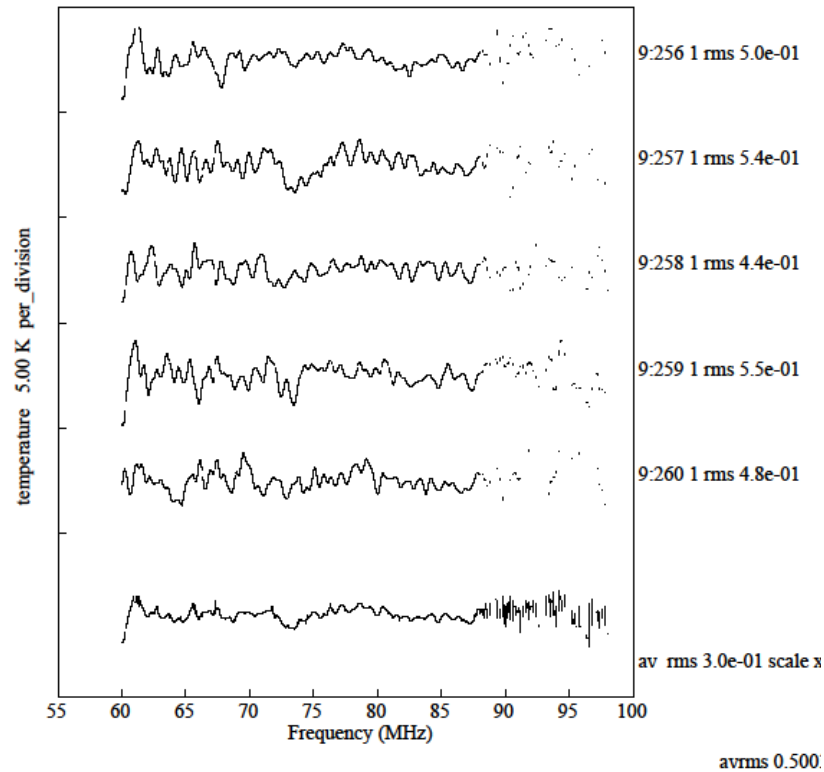


EDGES-3 test set-up in screen room at Haystack Observatory

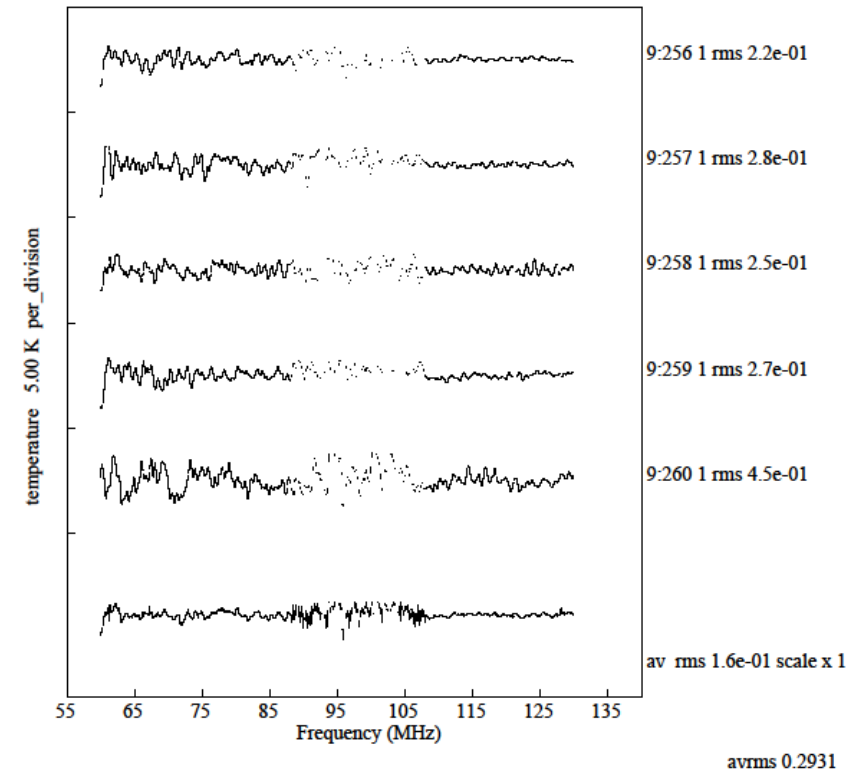
Test results from 3.5 hours each day on September 13,14,15 and 16



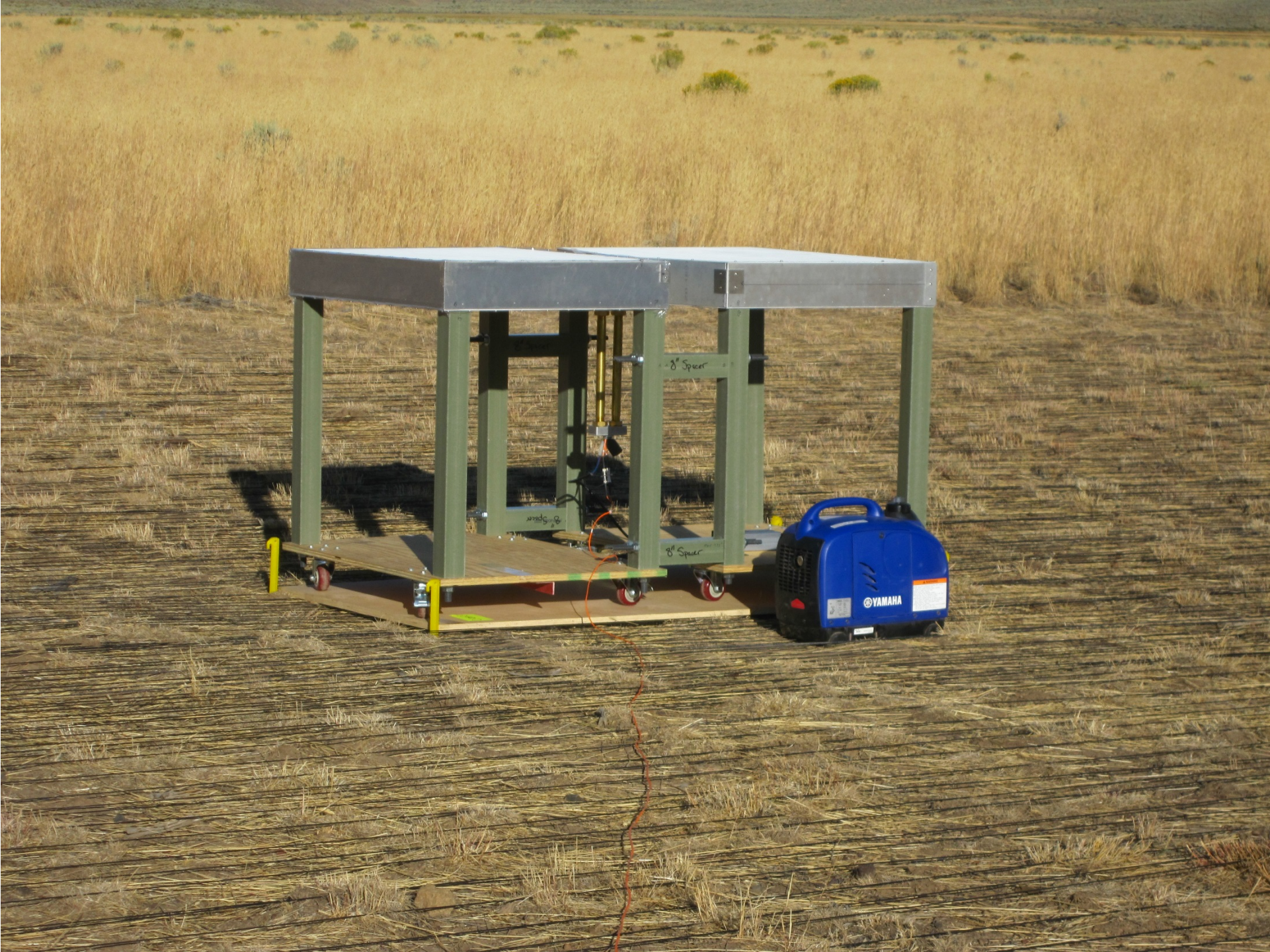
Spectrum without RFI excision

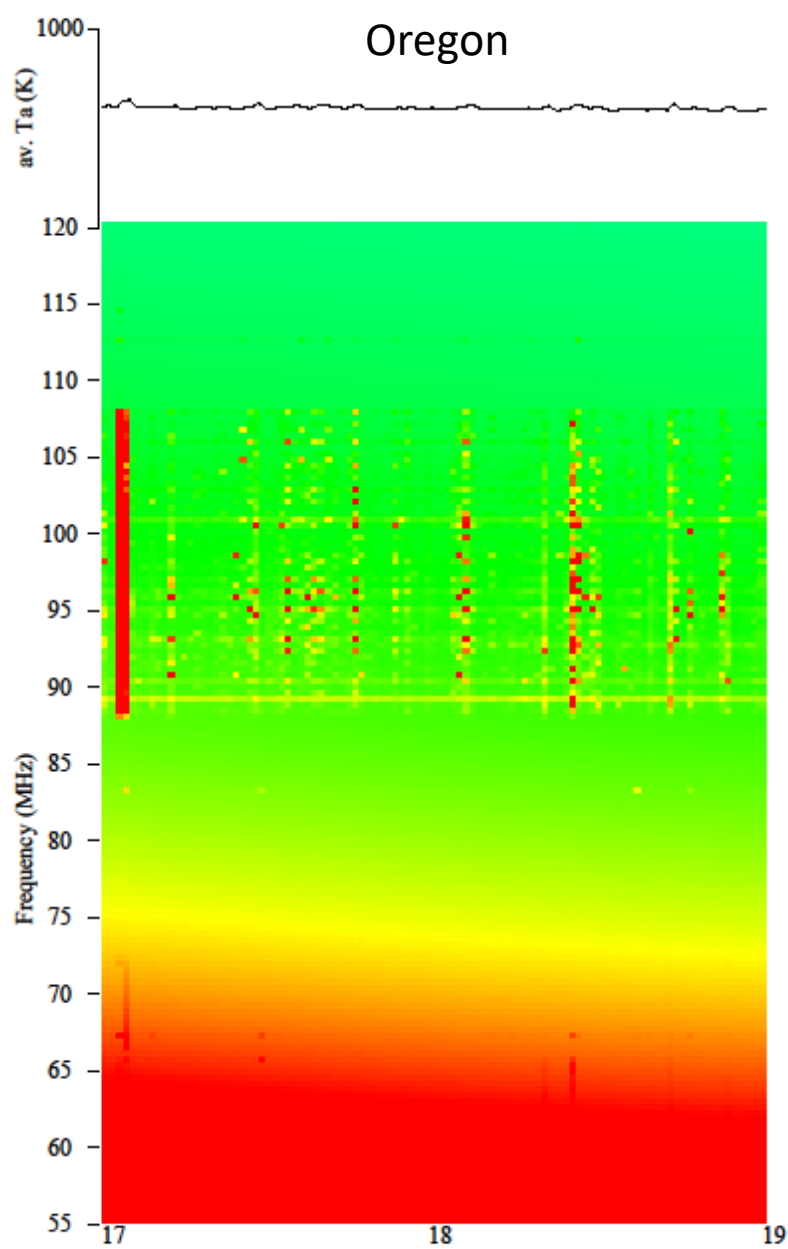


Spectra with RFI excision and 4 poly terms removed

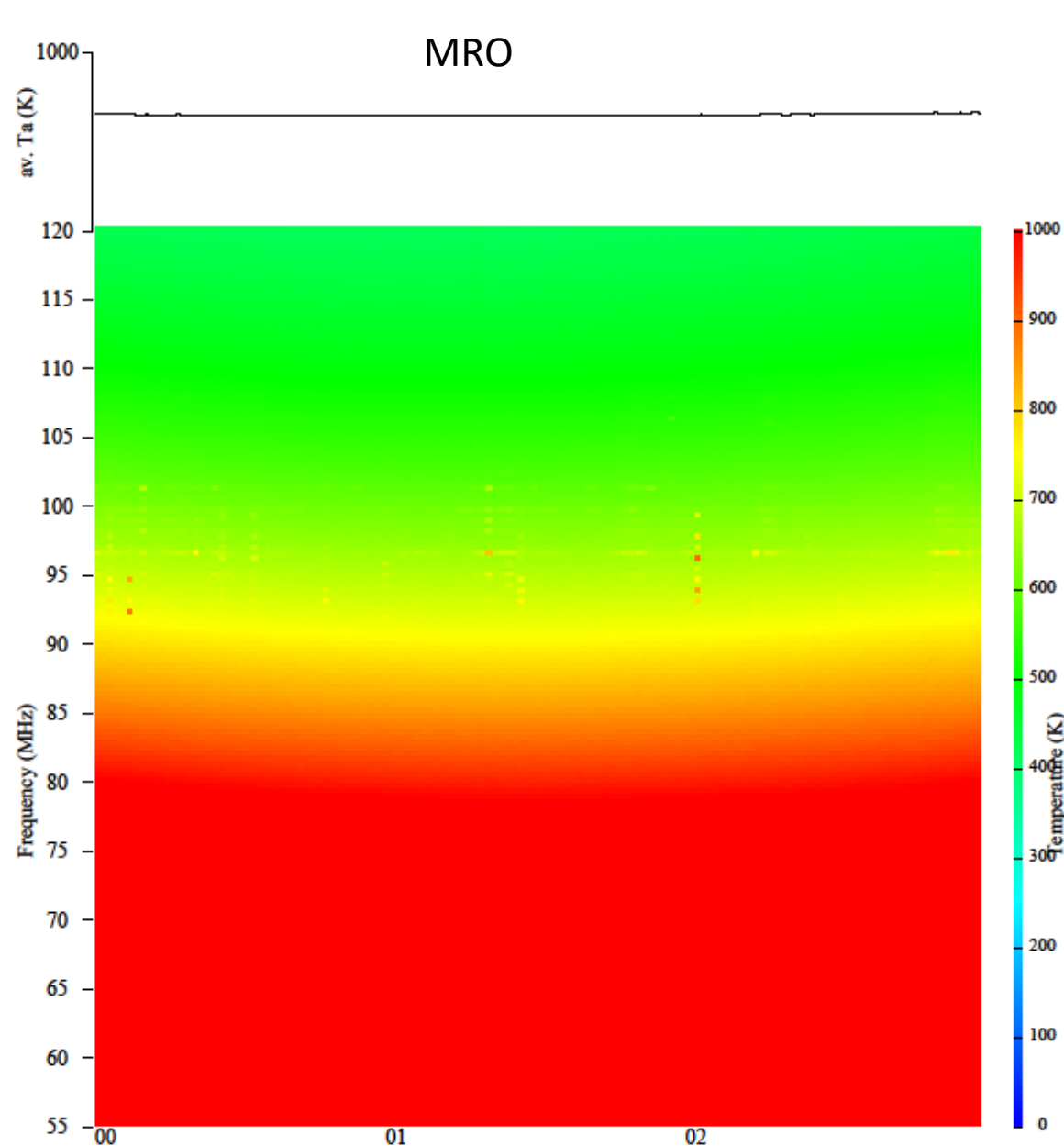


Spectra with RFI excision and 5 poly terms removed





UT 17 to 19
file: temp.acq
Mon Sep 23 15:05:07 2019
fstart 55 fstop 120 pfit 37 smooth 64 resol 391 kHz rfi 0.0 nline 307 secint 1962



UT 00 to 02
file: temp.acq
Mon Sep 23 16:26:31 2019
fstart 55 fstop 120 pfit 37 smooth 64 resol 391 kHz rfi 0.0 nline 412 secint 2633

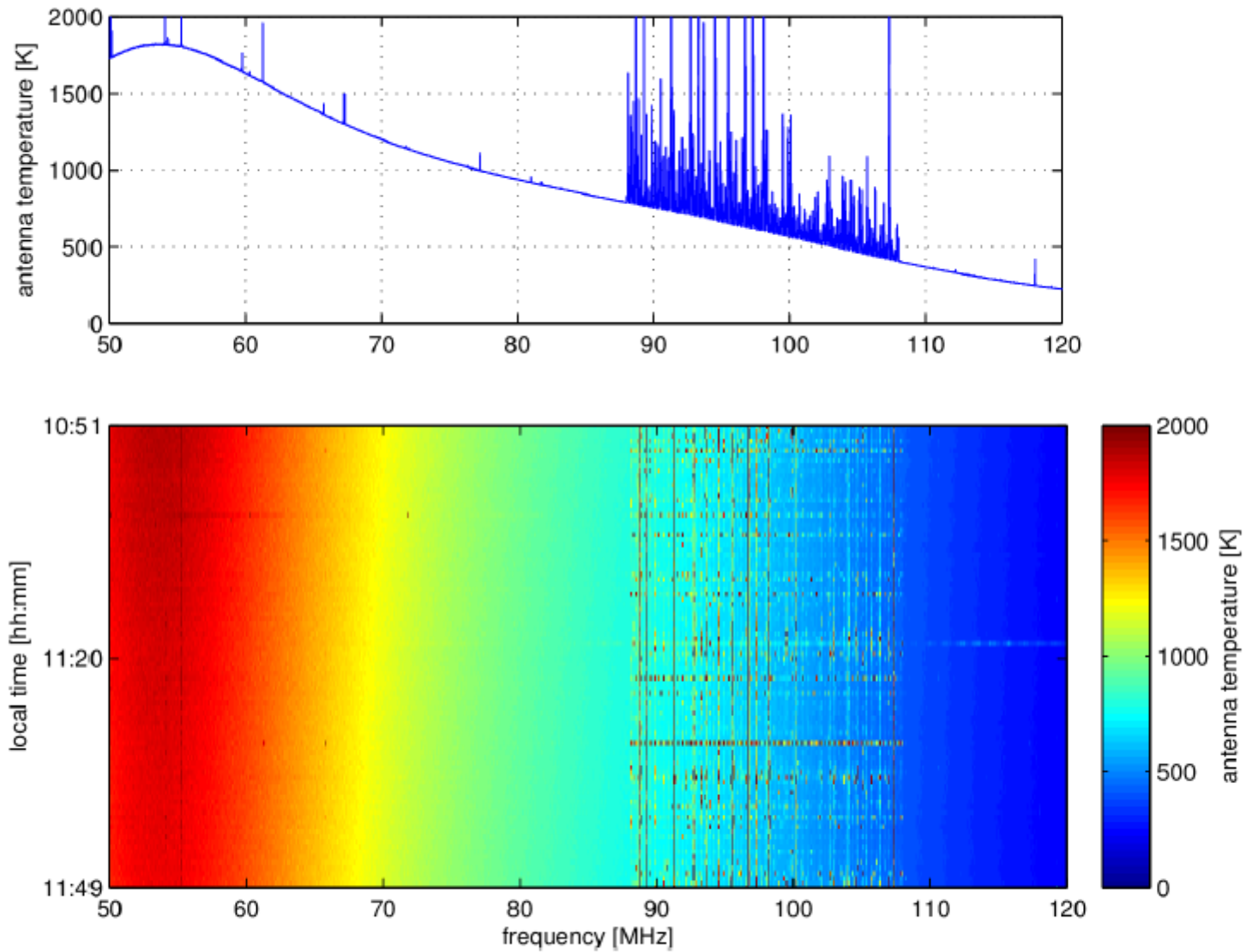
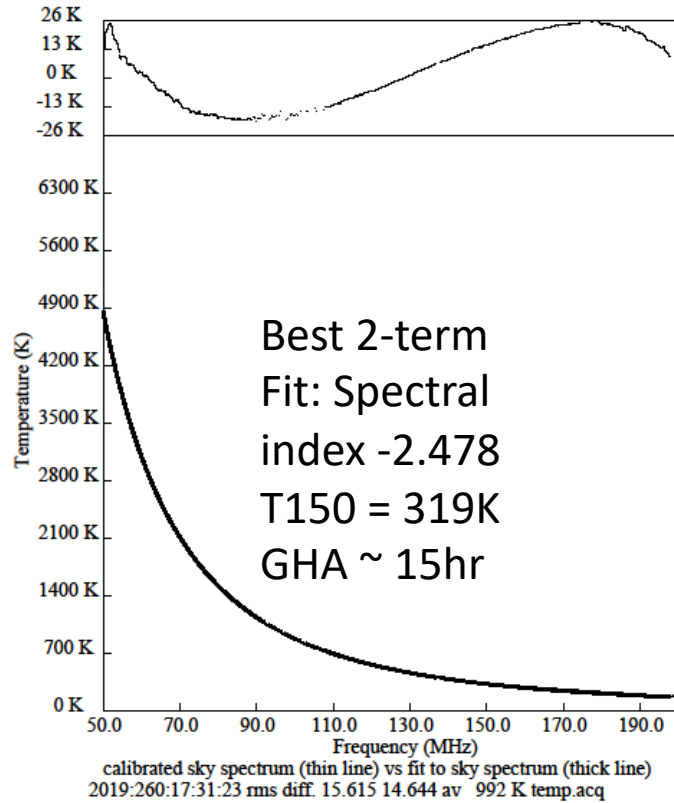
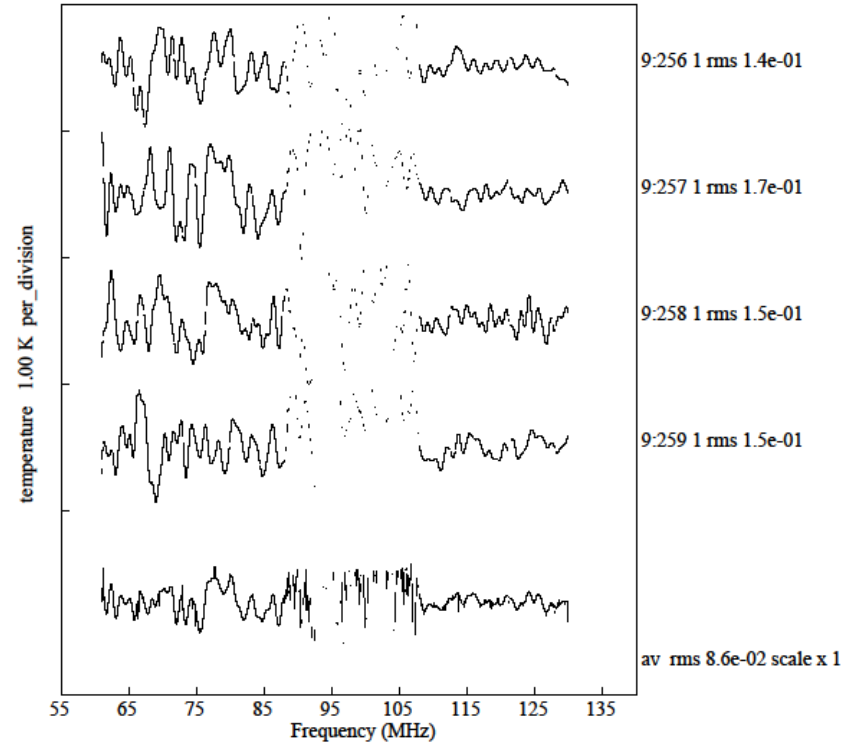


Figure 12: Site 1, low-band, NS orientation.
Comparison with a low-band test at Gund ranch Nevada
Raul Monsalve and Thomas Mozdzen July 2014

rms of 86 mK reached
with more smoothing and
RFI excision



Full calibrated spectrum
with 2-term fit to scale
and spectral index



avrms 0.1520

Calibrated spectrum 60 – 130
MHz 5-poly terms removed with
RFI excision and smoothing to
781 kHz. About 80% excised

Calibration and processing procedure

In the Lab:

- Measure S_{11} of the resistance the internal calibration load and its temperature coefficient
- Measure S -parms of the cable from the internal 8-position switch to the antenna input by manually placing known SOL calibration loads on antenna input
- Run automated 3-position switched spectra on internal ambient, hot, open and shorted cable and S_{11} data on internal SOL ambient, hot, open and shorted cable loads as well as LNA
- Calculate calibration of receiver LNA noise waves and receiver gain and offset
- Connect artificial antenna to receiver input and measure S_{11} and take 3-position switched spectra

In the field:

- Repeat automated calibration when needed
- Take 3-position switched spectra of the sky and antenna S_{11} data

Using EM simulations

- Estimate the antenna and ground plane loss

Obtain absolute calibration of sky spectrum processing

- Using calibration data, antenna S_{11} and losses obtain sky spectrum
- Use weighted least squares with up to 6 physics based “basis” functions to remove foreground, ionosphere and solve for hydrogen line signature
- Error estimates from covariance matrix MCMC etc.

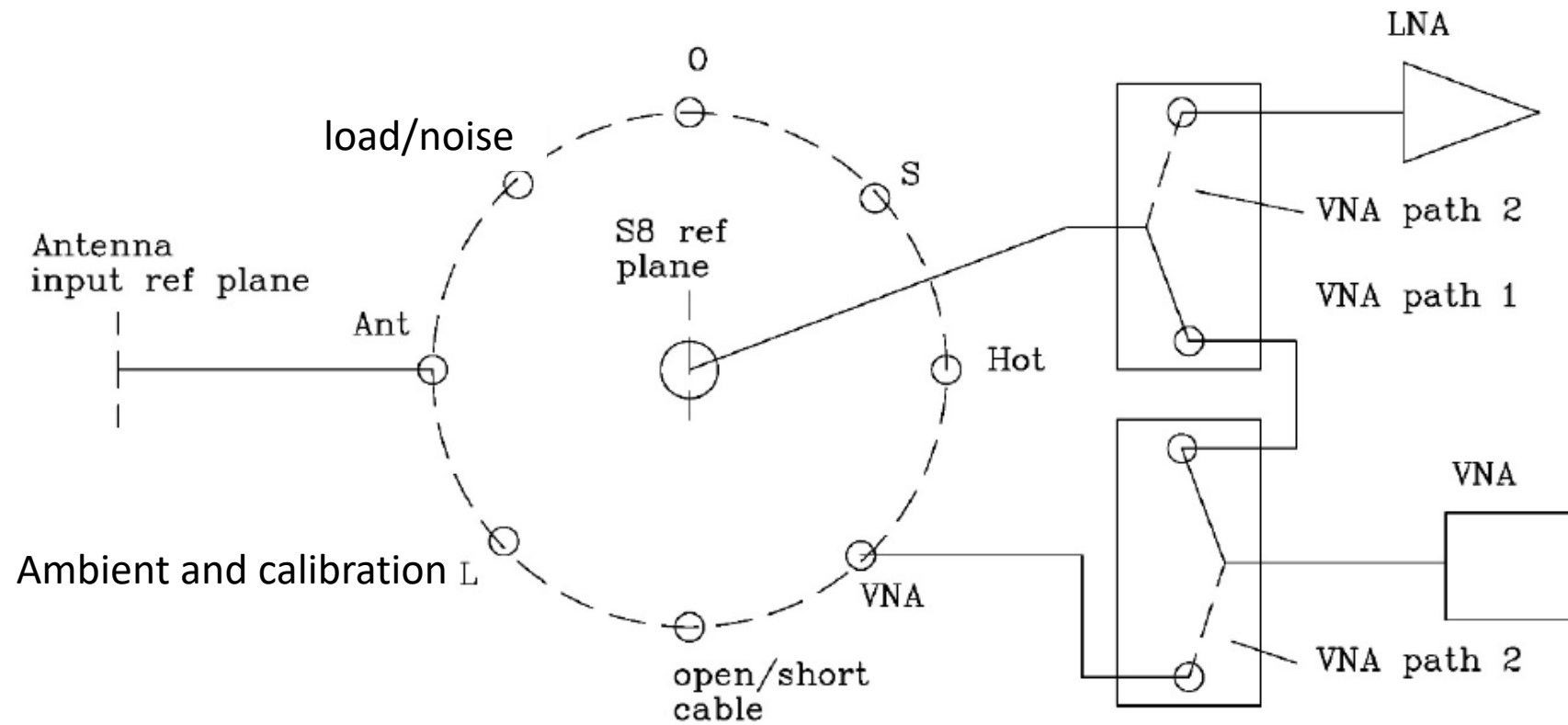
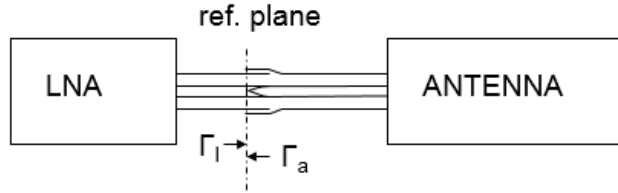


Figure 1. Schematic of change in reference plane.

S-parameters of input cable measured using calibration SOL placed on antenna input in Lab
 Loss \sim 0.07 dB or about 1.5 % at 75 MHz

Antenna to Low Noise Amplifier mismatch



Compensating for the antenna mismatch

$$T_{sky}(1 - |\Gamma|^2) = T_{sky}(1 - |\Gamma_a|^2)|F|^2$$

where Γ is the reflection from the LNA and

$$\Gamma = \frac{Z_a - Z_l^*}{Z_a + Z_l}$$

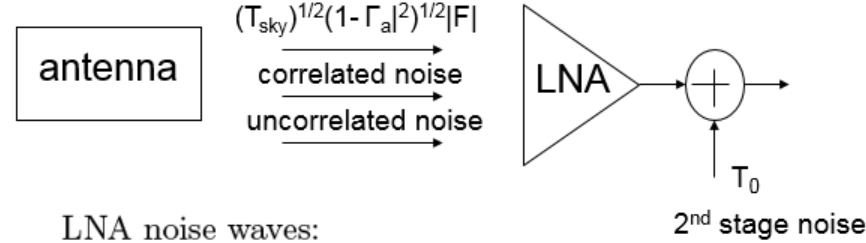
$$F = \frac{(1 - |\Gamma_l|^2)^{1/2}}{1 - \Gamma_a \Gamma_l}$$

where Γ_a and Γ_l are the reflections at 50 ohms ref. point

$$\Gamma_a = \frac{Z_a - 50}{Z_a + 50}$$

$$\Gamma_l = \frac{Z_l - 50}{Z_l + 50}$$

LNA noise waves reflected back from antenna



LNA noise waves:

$$T_{rec} = T_{sky}(1 - |\Gamma_a|^2)|F|^2 + T_u|\Gamma_a|^2|F|^2 + (T_c \cos(\phi) + T_s \sin(\phi))|\Gamma_a||F| + T_0$$

T_u is the uncorrelated wave

$T_c \cos(\phi)$ and $T_s \sin(\phi)$ are the correlated portions which depend on the phase, ϕ , of the reflected wave.

ϕ is the phase of $\Gamma_a F$

T_0 is the "second stage noise".

3 – position input switching – antenna, load, cal to take out "bandpass" and set temperature scale

$$P_{ant} = gT_{rec}$$

$$P_{load} = g(GT_{amb} + T_0)$$

$$P_{cal} = g(G(T_{amb} + T_{cal}) + T_0)$$

where g is the receiver gain and G is

$$G = 1 - |\Gamma_l|^2$$

T_{amb} is the ambient temperature and T_{cal} calibration noise

The calibrated receiver output, T_{3p} , is

$$\begin{aligned} T_{3p} &= \frac{T_{cal}(P_{ant} - P_{load})}{(P_{cal} - P_{load})} + T_{amb} \\ &= T_{sky}(1 - |\Gamma_a|^2)|F|^2 G^{-1} \\ &\quad + T_u|\Gamma_a|^2|F|^2 G^{-1} \\ &\quad + (T_c \cos(\phi) + T_s \sin(\phi))|\Gamma_a||F|G^{-1} \end{aligned}$$

Calibration requires measurements of antenna and LNA reflection coefficients as well correlated and uncorrelated LNA noise. Scale and offset are set from ambient and hot loads

Correction for losses

$$T = T_{\text{sky}}L + T_{\text{amb}}(1-L)$$

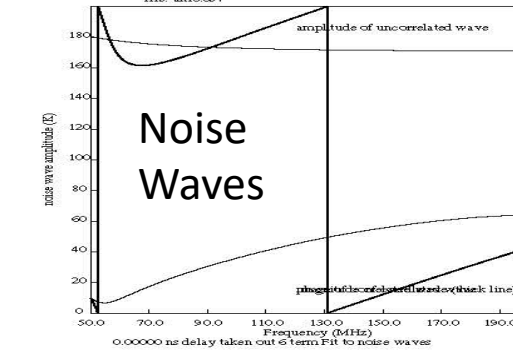
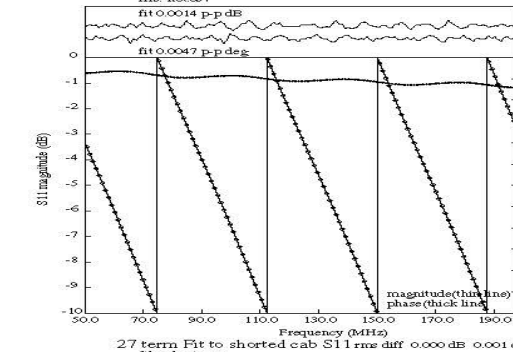
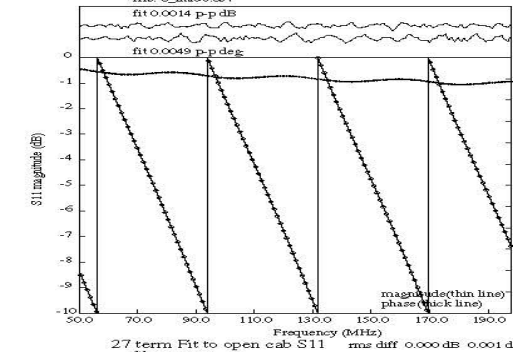
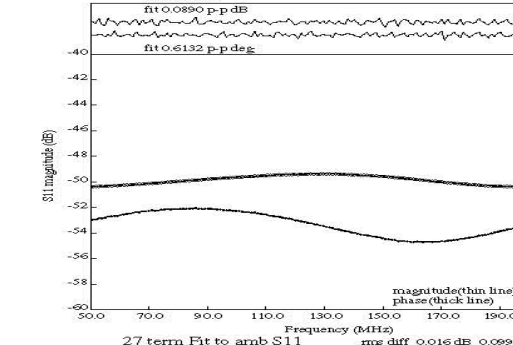
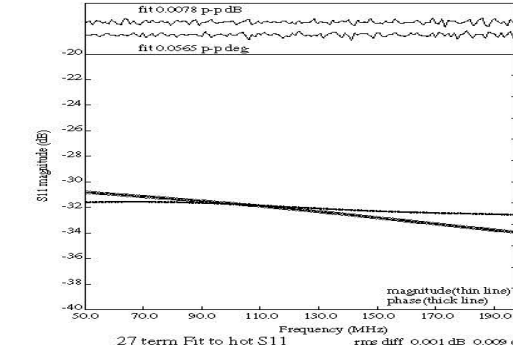
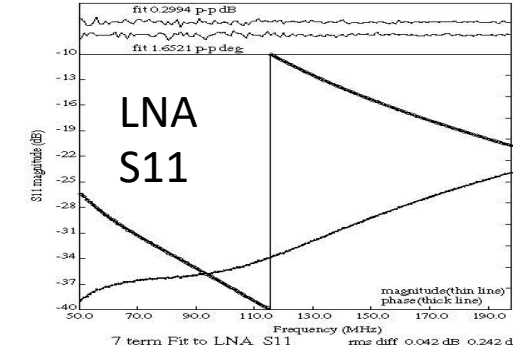
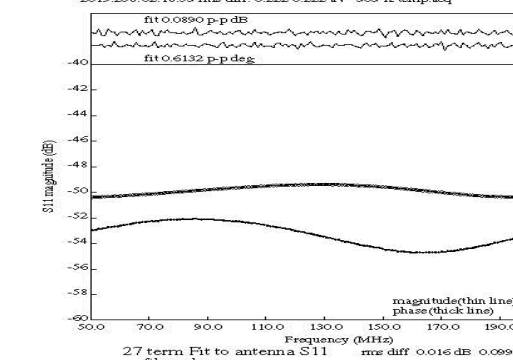
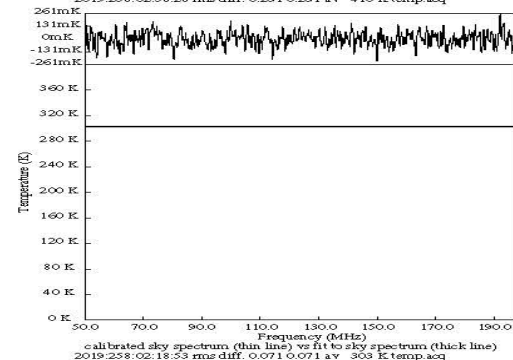
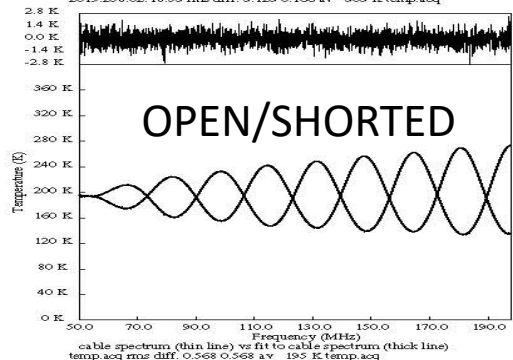
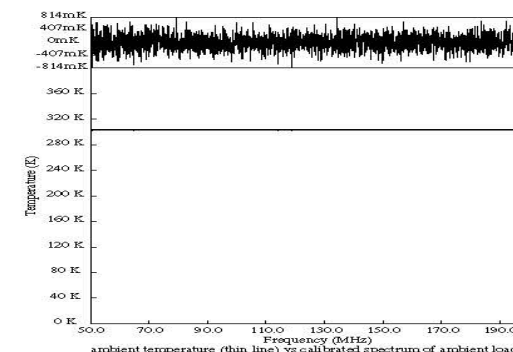
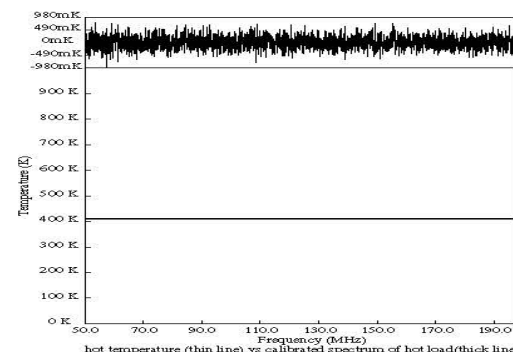
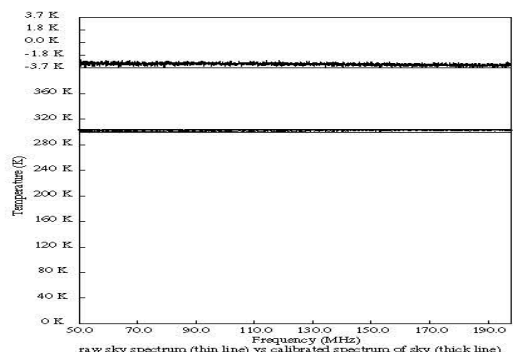
$$L = (1 - |\Gamma_a|^2)^{-1} |S_{21}|^2 (1 - |\Gamma|^2) / |1 - S_{22}\Gamma|^2$$

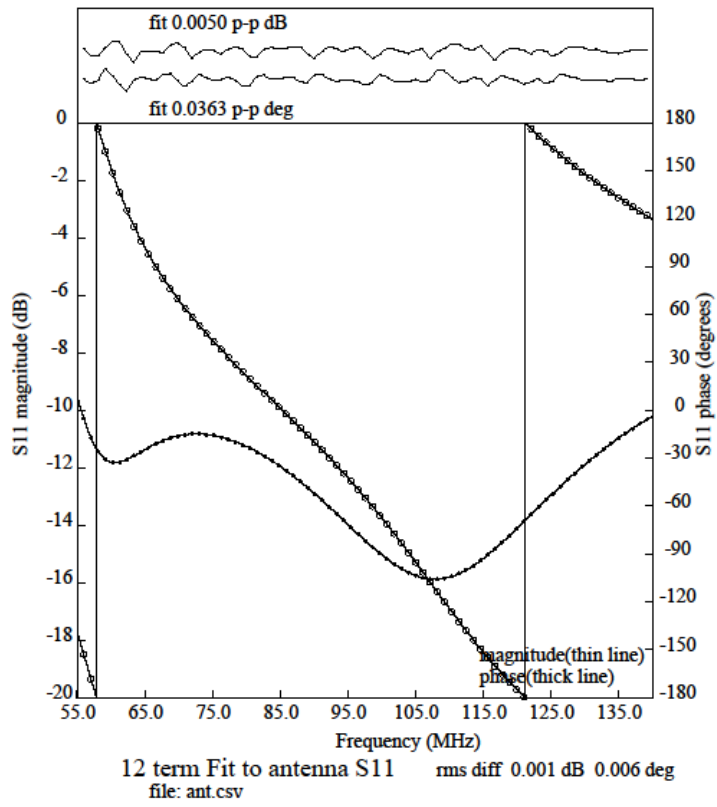
where:

Γ_a = reflection coefficient on antenna measured from reference plane at 8-position switch

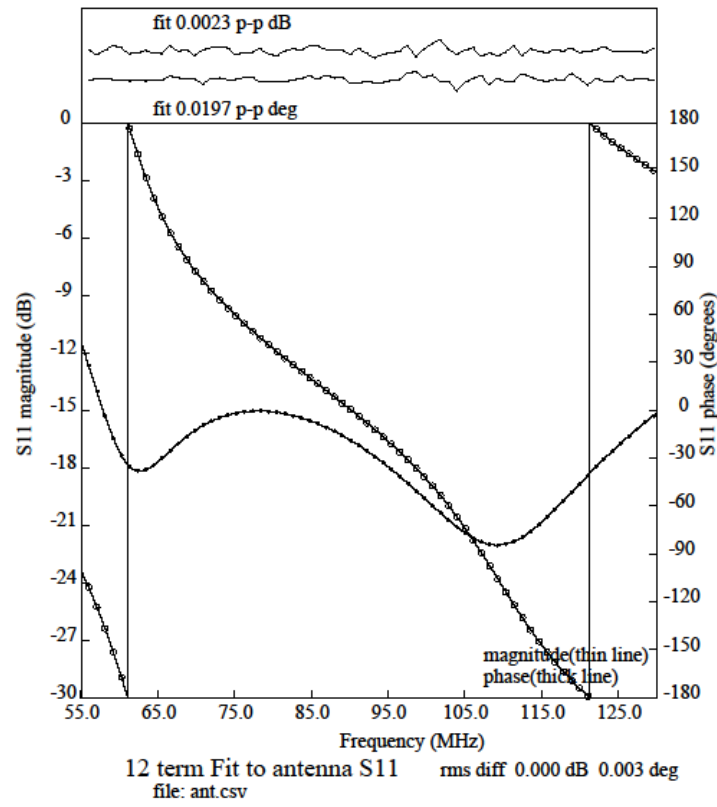
Γ = antenna reflection = $(\Gamma_a - S_{11}) / (S_{12}S_{21} - S_{11}S_{22} + S_{22}\Gamma_a)$

$S_{11}, S_{22}, S_{12}, S_{21}$ = scattering coefficients from reference plane to antenna connector on receiver input

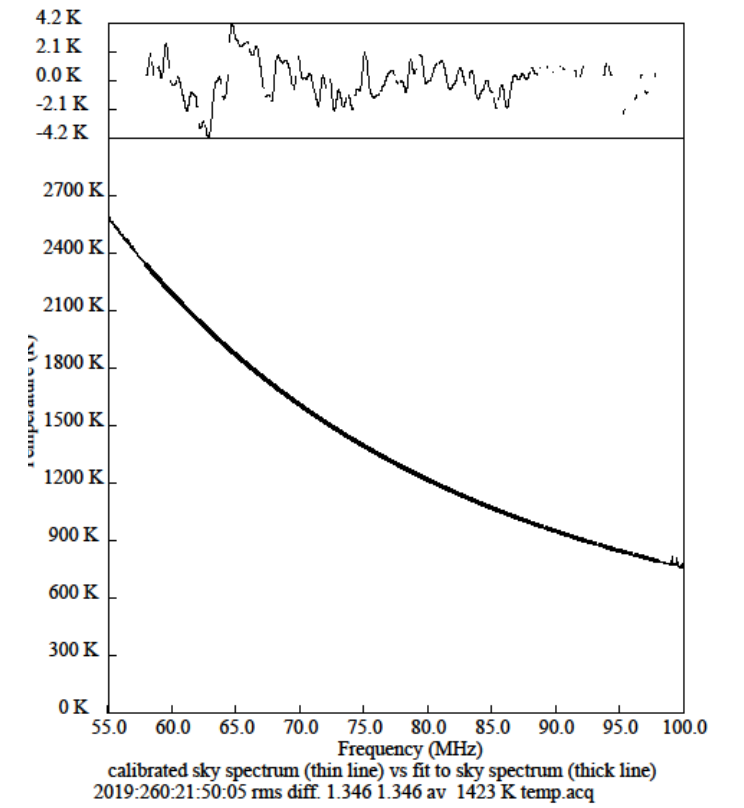




Antenna S11 on wire grid



on ground with
wire grid removed

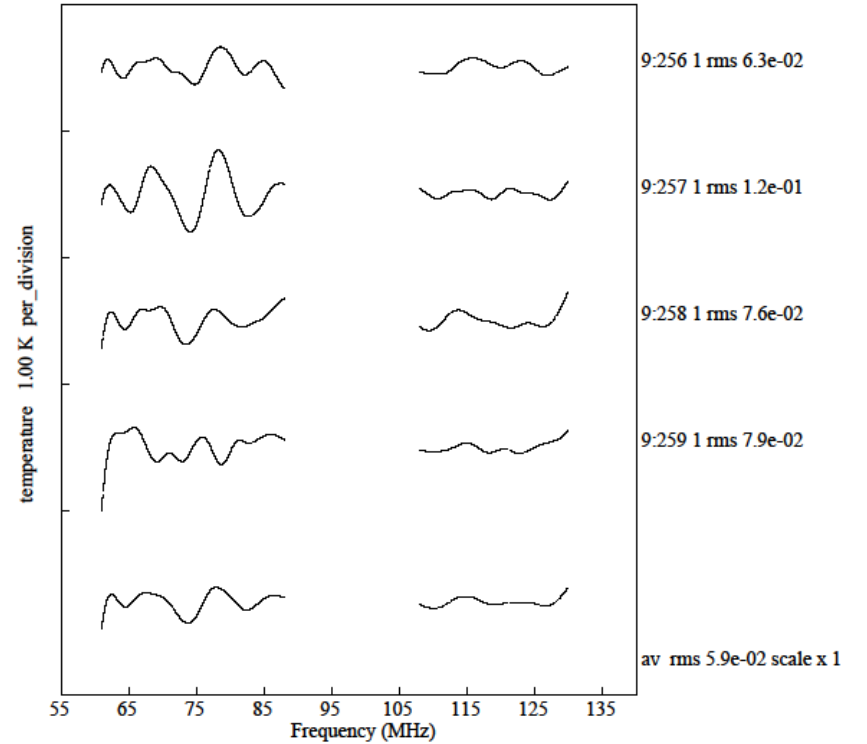
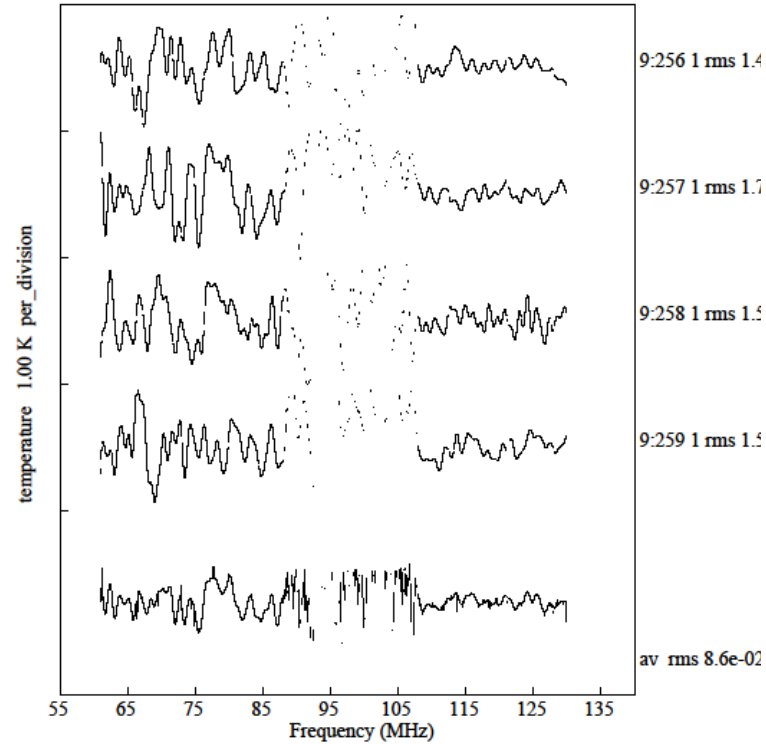
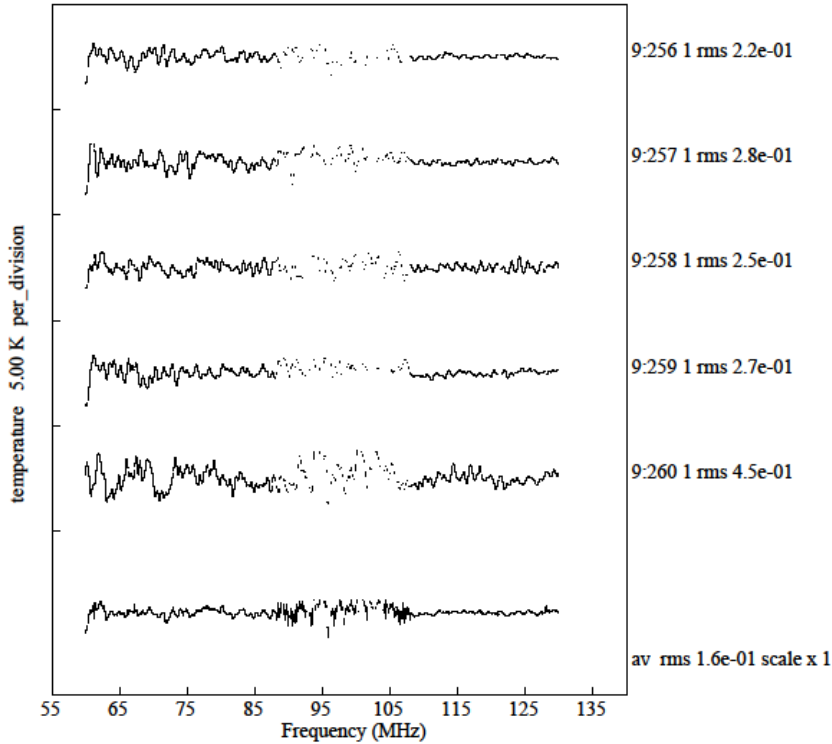


Tests of resolution and RFI excision

rms 160 mK

rms 86 mK

rms 59 mK



avrms 0.2931

avrms

avrms 0.0851

60-130 MHz 390 kHz

60-130 MHz 781 kHz

60-130 MHz 3MHz

5 poly terms removed

Thank you - Questions/Discussion

